METHOD FOR DETERMINING QUALITY INDICATORS OF ELECTRICAL POWER

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Abstract. This paper proposes to introduce a small number of quality indicators for electrical power in systems with three-phased consumers which are disbalanced, asymmetrical and distorting. These indicators are required for measuring and predetermining the effects of such consumers on the supply network in order to establish measures for improving the quality of power supply in the electrical networks.

Keywords: Quality indicators, electrical power, energy quality

1. Introduction

The problem related to the power quality in electrical networks are the following:

- frequency variations;
- active and reactive power shocks;
- distorting, asymmetrical and disbalanced regime.

This last problem has some very important consequences:

- the increase of power losses and active power;
- the interblocking of installations transport and supply capacity;
- difficulties related to the voltage regulating;
- parasitic torques in electrical machines;
- errors done by measure and control apparatus.

2. Three-phased disbalanced, asymmetrical and distorting consumer

Let it be a three-phased asymmetrical consumer R, S, T, whose impedances $Z_R \neq Z_S \neq Z_{T_c}$

Supplied from a three-phased voltage system $u_R, u_S, u_T \in u_{Rst}$ with disbalanced currents $i_R, i_S, i_T \in i_{RST}$ running under distorting conditions.

Instantaneous value expressions and the phase voltage u_{RST} and the i_{RST} currents are:

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$$u_{c(RST)} = \sum_{1}^{\infty} U_{k(RST)} \times \sqrt{2} \times \cos(k\omega t + \alpha_{k(RST)}) = U_{c(RST)} \times \cos\omega t - U_{s(RST)} \times \sin\omega t$$

$$u_{s(RST)} = \sum_{1}^{\infty} U_{k(RST)} \times \sqrt{2} \times \sin(k\omega t + \alpha_{k(RST)}) = U_{s(RST)} \times \cos \omega t + U_{c(RST)} \times \sin \omega t$$

$$i_{c(RST)} = \sum_{1}^{\infty} I_{k(RST)} \times \sqrt{2} \times \cos(k\omega t + \beta_{k(RST)}) = I_{c(RST)} \times \cos\omega t - I_{s(RST)} \times \sin\omega t$$
 (1)

$$i_{s(RST)} = \sum_{1}^{\infty} I_{k(RST)} \times \sqrt{2} \times \sin(k\omega t + \beta_{k(RST)}) = I_{s(RST)} \times \cos \omega t + I_{s(RST)} \times \sin \omega t$$

where:

$$U_{c(RST)} = \sum_{1}^{\infty} U_{k(RST)} \times \sqrt{2} \times \cos[\alpha_{k(RST)} + (k-1)\omega t]$$

$$U_{c(RST)} = \sum_{1}^{\infty} U_{k(RST)} \times \sqrt{2} \times \sin[\alpha_{k(RST)} + (k-1)\omega t]$$

$$I_{c(RST)} = \sum_{1}^{\infty} I_{k(RST)} \times \sqrt{2} \times \cos[\beta_{k(RST)} + (k-1)\omega t]$$

$$I_{s(RST)} = \sum_{1}^{\infty} I_{k(RST)} \times \sqrt{2} \times \sin[\beta_{k(RST)} + (k-1)\omega t]$$

Current and voltage systems satisfy the following requirements:

$$u_{CR} + u_{CS} + u_{CT} = u_{CO}$$

$$u_{SR} + u_{SS} + u_{ST} = u_{SO}$$

$$i_{CR} + i_{CS} + i_{CT} = i_{CO}$$

$$i_{SR} + i_{SS} + i_{ST} = i_{SO}$$
(3)

(2)

There has been demonstrated [1], [2], [3], that:

$$u_{RST}^2 = \frac{u_{C(RST)}^2 + u_{S(RST)}^2}{2}; i_{RST}^2 = \frac{i_{C(RST)}^2 + i_{S(RST)}^2}{2} (4)$$

$$p_{RST} = \frac{u_{C(RST)} \times i_{C(RST)} + u_{S(RST)} \times i_{S(RST)}}{2} \qquad q_{RST} = \frac{u_{C(RST)} \times i_{S(RST)} - u_{S(RST)} \times i_{C(RST)}}{2}$$
 (5)

$$s_{RST}^2 = u_{RST}^2 \times i_{RST}^2 \tag{6}$$

which represent characteristical functions of voltage u_{RST} , current i_{RST} , active power p_{RST} , complementary power c_{RST} , and apparent power s_{RST} . These functions of voltage and current are graphically represented in figure 1 and figure 2. Their property is that of being able to describe the variations of sizes characteristic to power quality as compared to the ideal conditions. The characteristic functions:

$$U_{RST}^{2} = \sum_{0}^{\infty} U_{k}^{2} = \frac{1}{T} \int_{0}^{T} u_{RST}^{2} \times dt = f \int_{0}^{\frac{1}{f}} u_{RST}^{2} \times dt$$
 (7)

$$I_{RST}^{2} = \sum_{0}^{\infty} I_{k}^{2} = \frac{1}{T} \int_{0}^{T} i_{RST}^{2} \times dt = f \int_{0}^{T} i_{RST}^{2} \times dt$$
 (8)

$$P_{RST} = \sum_{0}^{\infty} P_K = \frac{1}{T} \int_{0}^{T} p_{RST} \times dt = f \int_{0}^{T} p_{RST} \times dt$$
 (9)

$$Q_{RST} = \sum_{0}^{\infty} Q_K = \frac{1}{T} \int_{0}^{T} q_{RST} \times dt = f \int_{0}^{T} q_{RST} \times dt$$
 (10)

$$S_{RST}^2 = U_{RST}^2 \times I_{RST}^2 \tag{11}$$

enable us to obtain voltage, currents, active, inactive and apparent powers.

In them papers mentioned above these has been demonstrated that among the characteristic functions we have the relation:

$$s_{RST}^2 = p_{RST}^2 + q_{RST}^2 (12)$$

which represents the principle of power separation introduced by C. I. Budeanu. It also may by rendered as:

$$S^2 = P^2 + Q^2 + D^2 (13)$$

where D represents the distorting power.

3. Quality indicators

To define the power quality we do introduce two indicators (figure 1 and figure 2):

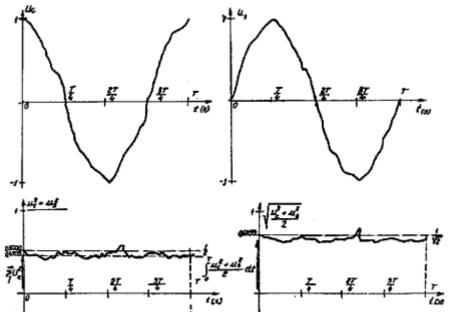


Fig. 1. The characteristic function of voltage.

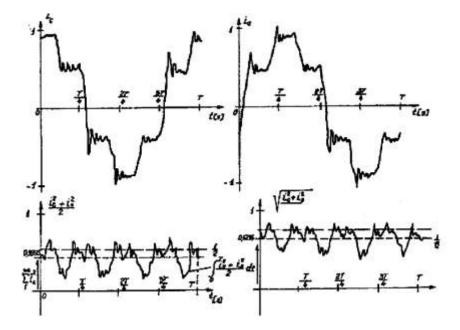


Fig. 2. The characteristic function of current.

load of the isolation: S_u

$$S_u = \int_0^T u_{RST}^2 \times dt = U_{RST}^2 \times T \tag{14}$$

• thermal load: Si

$$S_i = \int_0^T i_{RST}^2 \times dt = I_{RST}^2 \times T \tag{15}$$

These indicators are independent upon frequency variation and knid of sinusoidal energetical regime.

For sinusoidal energetical regime (S) there are two situations:

a) for frequency f = 50Hz, T = 1/f;

$$S_u(S,T) = \int_0^T u_{RS(RST)}^2 \times dt = U_{RS(RST)}^2 \times T$$
 (16)

$$Si(S,T) = \int_{0}^{T} i_{RS(RST)}^{2} \times dt = I_{RS(RST)}^{2} \times T$$
(17)

b) for frequency $f' \neq 50$ Hz, T = 1/f'

$$S_{u}(S,T') = \int_{0}^{T} u_{RST}^{2} \times dt = U_{RS(RST)}^{2} \times T'$$
 (18)

$$Si(S,T') = \int_{0}^{T} i_{RS(RST)}^{2} \times dt = I_{RS(RST)}^{2} \times T'$$
(19)

For non-sinusoidal energetical regime (N) there are also situations:

a) for frequency f = 50Hz, T = 1/f;

$$S_{u}(N,T) = \int_{0}^{T} u_{RN(RST0}^{2} \times dt = U_{RN(RST)}^{2} \times T$$
 (20)

$$Si(N,T) = \int_{0}^{T} i_{RN(RST)}^{2} \times dt = I_{RN(RST)}^{2} \times T$$
(21)

b) for frequency $f' \neq 50$ Hz, T = 1/f'

$$S_u(N,T') = \int_0^T u_{RST}^2 \times dt = U_{RN(RST)}^2 \times T'$$
 (22)

$$Si(N,T') = \int_{0}^{T} i_{RN(RST)}^{2} \times dt = I_{RN(RST)}^{2} \times T'$$
(23)

By using the representations in relative units of functions $u_{RST}^2(t)$ and $i_{RST}^2(t)$ in figure 1 and figure 2, we can define these function such as:

$$S_{u}(S,T) = U^{2}T = \frac{1}{2} \times T$$

$$S_{i}(S,T) = I^{2}T = \frac{1}{2} \times T$$
(24)

$$S_u(S,T') = U^2 T' = \frac{1}{2} \times T'$$

 $S_i(S,T') = I^2 T' = \frac{1}{2} \times T'$
(25)

$$S_{u}(N,T) = \sum_{k=1}^{n} U_{k}^{2}T = 0,4709T$$

$$S_{i}(N,T) = \sum_{k=1}^{n} I_{k}^{2}T = 0,3995T$$
(26)

$$S_{u}(N,T') = \sum_{1}^{n} U_{k}^{2} T' = 0,4709T'$$

$$S_{i}(N,T') = \sum_{1}^{n} I_{k}^{2} T' = 0,3995T'$$
(27)

Having as reference basis the functions $S_u(S,T)$ and $S_i(S,T)$ corresponding to the distorting energetical regime at frequency f=Hz, any non-sinusoidal energetical regime at frequency $f'\neq f=50Hz$ is characterized by functions S_u and S_i .

The deviation from energy quality (voltage and current) can be defined by following functions:

$$\Delta S_{u}(N,T') = \frac{S_{u}(N,T') - S_{u}(S,T)}{S_{u}(N,T')} = 1 - \frac{U^{2}T}{\sum_{k=1}^{n} U_{k}^{2}T'}$$
(28)

$$\Delta S_i(N,T') = \frac{S_i(N,T') - S_i(S,T)}{Si(N,T')} = 1 - \frac{I^2 T}{\sum_{k=1}^{n} I_k^2 T'}$$
(29)

The quality deviation due to the network frequency variation is defined by the functions:

$$\Delta S_u(S, T') = \frac{S_u(S, T') - S_u(S, T)}{S_u(S, T')} = 1 - \frac{T}{T'}$$
(30)

$$\Delta S_i(S, T') = \frac{S_i(S, T') - S_i(S, T)}{S_i(S, T')} = 1 - \frac{T}{T'}$$
(31)

and then we have the relations:

$$\Delta S(f) = 1 - \frac{T}{T'} = 1 - \frac{f'}{f} = \Delta S_u(S, T') = \Delta S_i(S, T')$$
 (32)

The quality deviation due to non-sinusoidal energetical regime is defined by the following functions:

$$\Delta S_u(D,T) = \frac{S_u(N,T) - S_u(S,T)}{S_u(N,T)} = 1 - \frac{U^2}{\sum_{k=0}^{n} U_k^2}$$
(33)

$$\Delta S_u(D,T') = \frac{S_u(N,T) - S_u(S,T')}{S_u(N,T')} = 1 - \frac{U^2}{\sum_{k=1}^n U_k^2}$$
(34)

$$\Delta S_i(D,T) = \frac{S_i(N,T) - S_i(S,T)}{S_u(N,T)} = 1 - \frac{I^2}{\sum_{k=1}^{n} I_k^2}$$
(35)

$$\Delta S_i(D, T') = \frac{S_i(N, T') - S_i(S, T')}{S_u(N, T')} = 1 - \frac{I^2}{\sum_{k=1}^{n} I_k^2}$$
(36)

$$\Delta S_u(D) = S_u(D,T) = \Delta S_u(D,T'); \tag{37}$$

$$\Delta S_i(D) = S_i(D,T) = \Delta S_i(D,T); \tag{38}$$

Under these conditions, the deviation from energy quality (voltage and current) can be defined as follows:

$$\Delta S_{u}(N,T') = 1 - [1 - \Delta S_{u}(D)] \times [1 - \Delta S(f)]$$
(39)

$$\Delta S_i(N,T') = 1 - [1 - \Delta S_i(D)] \times [1 - \Delta S(f)] \tag{40}$$

These should be noted that from the representations of figure 1 and 2, when it is operated at a frequency, f = 47Hz, it results a deviation from energy quality $\Delta S_{ii}(N,T') = 12\%$ and $\Delta S_{ii}(N,T') = 24,8\%$.

A unique energy quality indicator would be that which is including both insulation load and thermal load.

Laying down the characteristic $S_{ui(RST)} = (\frac{u_{c(RST)}^2 + u_{s(RST)}^2}{2})(\frac{i_{c(RST)}^2 + i_{c(RST)}^2}{2})$ for an electrical period T or T', we obtain the characteristic of <u>insulations load and thermal load</u>, SIT.

For sinusoidal energetical regime at frequency f = 50Hz, T = 1/f, we have:

$$SIT(S,T) = \int_{s_{1}}^{T} \left(\frac{u_{c(RST)}^{2} + u_{s(RST)}^{2}}{2}\right) \left(\frac{i_{c(RST)}^{2} + i_{c(RST)}^{2}}{2}\right) dt = U^{2}I^{2}T$$
(41)

While for non-sinusoidal energetical regime at frequency $f' \neq f = 50$ Hz, T' = 1/f', we have:

$$SIT(N,T') = \int_{0}^{T'} \left(\frac{u_{c(RST)}^2 + u_{s(RST)}^2}{2}\right) \left(\frac{i_{c(RST)}^2 + i_{c(RST)}^2}{2}\right) dt = \sum_{k=1}^{n} U_k^2 \sum_{k=1}^{n} I_k^2 T'$$
(42)

The deviation of isolation and thermal load characteristic is:

$$\Delta SIT(N,T') = \frac{SIT(N,T') - SIT(S,T)}{SIT(N,T')} = 1 - [1 - \Delta S_u(N,T')][1 - \Delta S_i(N,T')]\frac{T'}{T}$$
(43)

Conclusions

There have been suggested the indications for isolation load and thermal load to characterize the derivation from the electric power quality. With their help we can define a general energy quality indicator by means of the isolation and thermal load deviation. The main advantage is that these indicators can be determined without performing any harmonic expression.

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